

# ECE 604, Lecture 9

Mon, Jan 28, 2019

## Contents

<b>1 Gyrotropic Media</b>	<b>2</b>
<b>2 Wave Polarization</b>	<b>4</b>
2.1 Arbitrary Polarization Case and Axial Ratio . . . . .	7
<b>3 Polarization and Power Flow</b>	<b>9</b>

## 1 Gyrotropic Media

This section presents deriving the permittivity tensor of a gyrotropic medium in the ionosphere. Our ionosphere is always biased by a static magnetic field due to the Earth's magnetic field. But in this derivation, one assumes that the ionosphere has a static magnetic field polarized in the  $z$  direction, namely that  $\mathbf{B} = \hat{z}B_0$ . Now, the equation of motion from the Lorentz force law for an electron with  $q = -e$ , becomes

$$m_e \frac{d\mathbf{v}}{dt} = -e(\mathbf{E} + \mathbf{v} \times \mathbf{B}) \quad (1.1)$$

Next, let us assume that the electric field is polarized in the  $xy$  plane. Also,  $\mathbf{v} = d\mathbf{r}/dt$ , and in the frequency domain, the above becomes

$$m_e \omega^2 x = e(E_x + j\omega B_0 y) \quad (1.2)$$

$$m_e \omega^2 y = e(E_y - j\omega B_0 x) \quad (1.3)$$

The above equations cannot be solved easily for  $x$  and  $y$  in terms of the electric field because they correspond to a two-by-two matrix system with cross coupling between the unknowns  $x$  and  $y$ . But they can be simplified as follows. We can multiply (1.3) by  $\pm j$  and add it to (1.2) to get two decoupled equations:

$$m_e \omega^2 (x + jy) = e[(E_x + jE_y) + \omega B_0 (x + jy)] \quad (1.4)$$

$$m_e \omega^2 (x - jy) = e[(E_x - jE_y) - \omega B_0 (x - jy)] \quad (1.5)$$

Defining new variables such that

$$s_{\pm} = x \pm jy \quad (1.6)$$

$$E_{\pm} = E_x \pm jE_y \quad (1.7)$$

then (1.4) and (1.5) become

$$m_e \omega^2 s_{\pm} = e(E_{\pm} \pm \omega B_0 s_{\pm}) \quad (1.8)$$

Thus, solving the above yields

$$s_{\pm} = \frac{e}{m_e \omega^2 \mp e B_0 \omega} E_{\pm} = C_{\pm} E_{\pm} \quad (1.9)$$

where

$$C_{\pm} = \frac{e}{m_e \omega^2 \mp e B_0 \omega} \quad (1.10)$$

By this manipulation, the above equations (1.2) and (1.3) transform to new equations where there is no cross coupling between  $s_{\pm}$  and  $E_{\pm}$ . The mathematical parlance for this is the diagonalization of a matrix equation. Thus, the new equation can be solved easily.

Next, one can define  $P_x = -Nex$ ,  $P_y = -Ney$ , and that  $P_{\pm} = P_x \pm jP_y = -Nes_{\pm}$ , then it can be shown that

$$P_{\pm} = \varepsilon_0 \chi_{\pm} E_{\pm} \quad (1.11)$$

The expression for  $\chi_{\pm}$  can be derived, or

$$\chi_{\pm} = -\frac{NeC_{\pm}}{\varepsilon_0} = -\frac{Ne}{\varepsilon_0} \frac{e}{m_e \omega^2 \mp eB_0 \omega} = -\frac{\omega_p^2}{\omega^2 \mp \Omega \omega} \quad (1.12)$$

where  $\Omega$  and  $\omega_p$  are the cyclotron frequency<sup>1</sup> and plasma frequency, respectively.

$$\Omega = \frac{eB_0}{m_e}, \quad \omega_p^2 = \frac{Ne^2}{m_e \varepsilon_0} \quad (1.13)$$

At the cyclotron frequency, a solution exists to the equation of motion (1.1) without a forcing function, which in this case is the electric field. Thus, at this frequency, the solution blows up with a forcing function,  $E_{\pm}$  not being zero. This is like what happens to an LC tank circuit at resonance.

Now, one can rewrite (1.11) in terms of the original variables  $P_x$ ,  $P_y$ ,  $E_x$ ,  $E_y$ , or

$$\begin{aligned} P_x &= \frac{P_+ + P_-}{2} = \frac{\varepsilon_0}{2} (\chi_+ E_+ + \chi_- E_-) = \frac{\varepsilon_0}{2} [\chi_+ (E_x + jE_y) + \chi_- (E_x - jE_y)] \\ &= \frac{\varepsilon_0}{2} [(\chi_+ + \chi_-) E_x + j(\chi_+ - \chi_-) E_y] \end{aligned} \quad (1.14)$$

$$\begin{aligned} P_y &= \frac{P_+ - P_-}{2j} = \frac{\varepsilon_0}{2j} (\chi_+ E_+ - \chi_- E_-) = \frac{\varepsilon_0}{2j} [\chi_+ (E_x + jE_y) - \chi_- (E_x - jE_y)] \\ &= \frac{\varepsilon_0}{2j} [(\chi_+ - \chi_-) E_x + j(\chi_+ + \chi_-) E_y] \end{aligned} \quad (1.15)$$

The above relationship has to be expressed using a tensor where

$$\mathbf{P} = \varepsilon_0 \bar{\boldsymbol{\chi}} \cdot \mathbf{E} \quad (1.16)$$

where  $\mathbf{P} = [P_x, P_y]$ , and  $\mathbf{E} = [E_x, E_y]$ . From the above,  $\bar{\boldsymbol{\chi}}$  is of the form

$$\bar{\boldsymbol{\chi}} = \frac{1}{2} \begin{pmatrix} (\chi_+ + \chi_-) & j(\chi_+ - \chi_-) \\ -j(\chi_+ - \chi_-) & (\chi_+ + \chi_-) \end{pmatrix} = \begin{pmatrix} -\frac{\omega_p^2}{\omega^2 - \Omega^2} & -j\frac{\omega_p^2 \Omega}{\omega(\omega^2 - \Omega^2)} \\ j\frac{\omega_p^2 \Omega}{\omega(\omega^2 - \Omega^2)} & -\frac{\omega_p^2}{\omega^2 - \Omega^2} \end{pmatrix} \quad (1.17)$$

Notice that in the above, when the  $\mathbf{B}$  is turned off or  $\Omega = 0$ , the above resembles the solution of a collisionless, cold plasma.

For the electric field in the  $z$  direction, it is unaffected by the magnetic field as can be seen from the Lorentz force law or (1.1). Hence, it behaves like a

<sup>1</sup>This is also called the gyrofrequency.

simple collisionless plasma without a biasing magnetic field. Consequently, the above can be generalized to 3D to give

$$\bar{\chi} = \begin{bmatrix} \chi_0 & j\chi_1 & 0 \\ -j\chi_1 & \chi_0 & 0 \\ 0 & 0 & \chi_p \end{bmatrix} \quad (1.18)$$

where  $\chi_p = -\omega_p^2/\omega^2$ .

Using the fact that  $\mathbf{D} = \varepsilon_0\mathbf{E} + \mathbf{P} = \varepsilon_0(\bar{\mathbf{I}} + \bar{\chi}) \cdot \mathbf{E} = \bar{\varepsilon} \cdot \mathbf{E}$ , the above implies that

$$\bar{\varepsilon} = \varepsilon_0 \begin{bmatrix} 1 + \chi_0 & j\chi_1 & 0 \\ -j\chi_1 & 1 + \chi_0 & 0 \\ 0 & 0 & 1 + \chi_p \end{bmatrix} \quad (1.19)$$

Please notice that the above tensor is a hermitian tensor. We shall learn later that this is the hallmark of a lossless medium.

Another characteristic of a gyrotropic medium is that a linearly polarized wave will rotate when passing through it. This is the Faraday rotation effect, which we shall learn later. This phenomenon poses a severe problem to Earth-to-satellite communication using linearly polarized wave as it requires the alignment of the Earth-to-satellite antennas. This can be avoided using a rotatingly polarized wave, called a circularly polarized wave that we shall learn in the next section.

## 2 Wave Polarization

Studying wave polarization is very important for communication purposes. A wave whose electric field is pointing in the  $x$  direction while propagating in the  $z$  direction is a linearly polarized (LP) wave. The same can be said of one with electric field polarized in the  $y$  direction. It turns out that a linearly polarized wave suffers from Faraday rotation when it propagates through the ionosphere. For instance, an  $x$  polarized wave can become a  $y$  polarized due to Faraday rotation. So its polarization becomes ambiguous: to overcome this, Earth to satellite communication is done with circularly polarized (CP) waves. So even if the electric field vector is rotated by Faraday's rotation, it remains to be a CP wave. We will study these polarized waves next.

We can write a general uniform plane wave propagating in the  $z$  direction as

$$\mathbf{E} = \hat{x}E_x(z, t) + \hat{y}E_y(z, t) \quad (2.1)$$

Clearly,  $\nabla \cdot \mathbf{E} = 0$ , and  $E_x(z, t)$  and  $E_y(z, t)$  are solutions to the one-dimensional wave equation. For a time harmonic field, the two components may not be in phase, and we have in general

$$E_x(z, t) = E_1 \cos(\omega t - \beta z) \quad (2.2)$$

$$E_y(z, t) = E_2 \cos(\omega t - \beta z + \alpha) \quad (2.3)$$

where  $\alpha$  denotes the phase difference between these two waves components. We shall study how the linear superposition of these two components behaves for different  $\alpha$ 's. First, we set  $z = 0$  to observe this field. Then

$$\mathbf{E} = \hat{x}E_1 \cos(\omega t) + \hat{y}E_2 \cos(\omega t + \alpha) \quad (2.4)$$

For  $\alpha = \frac{\pi}{2}$

$$E_x = E_1 \cos(\omega t), \quad E_y = E_2 \cos(\omega t + \pi/2) \quad (2.5)$$

Next, we evaluate the above for different  $\omega t$ 's

$$\omega t = 0, \quad E_x = E_1, \quad E_y = 0 \quad (2.6)$$

$$\omega t = \pi/4, \quad E_x = E_1/\sqrt{2}, \quad E_y = -E_2/\sqrt{2} \quad (2.7)$$

$$\omega t = \pi/2, \quad E_x = 0, \quad E_y = -E_2 \quad (2.8)$$

$$\omega t = 3\pi/4, \quad E_x = -E_1/\sqrt{2}, \quad E_y = -E_2/\sqrt{2} \quad (2.9)$$

$$\omega t = \pi, \quad E_x = -E_1, \quad E_y = 0 \quad (2.10)$$

The tip of the vector field  $\mathbf{E}$  traces out an ellipse as show in Figure 1. With the thumb pointing in the  $z$  direction, and the wave rotating in the direction of the fingers, such a wave is called left-hand elliptically polarized (LHEP) wave.

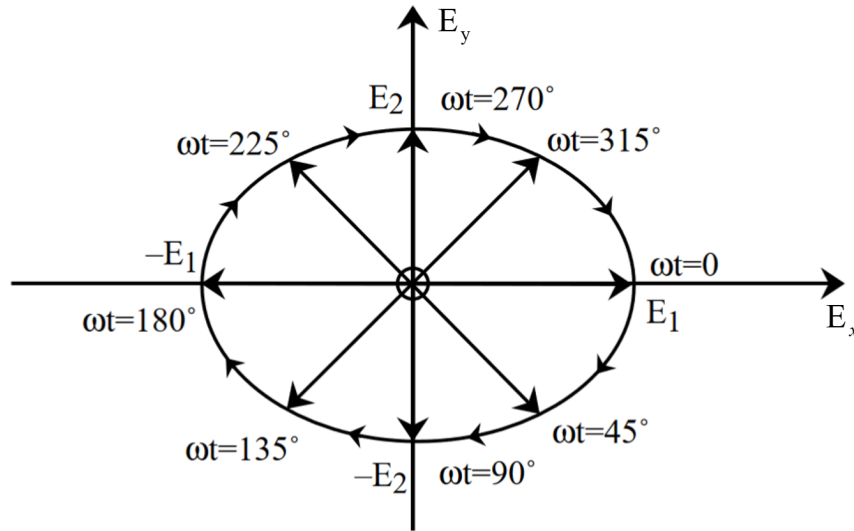


Figure 1:

When  $E_1 = E_2$ , the ellipse becomes a circle, and we have a left-hand circularly polarized (LHCP) wave. When  $\alpha = -\pi/2$ , the wave rotates in the

counter-clockwise direction, and the wave is either right-hand elliptically polarized (RHEP), or right-hand circularly polarized (RHCP) wave depending on the ratio of  $E_1/E_2$ . Figure 2 shows the different polarizations of the wave for different phase differences and amplitude ratio.

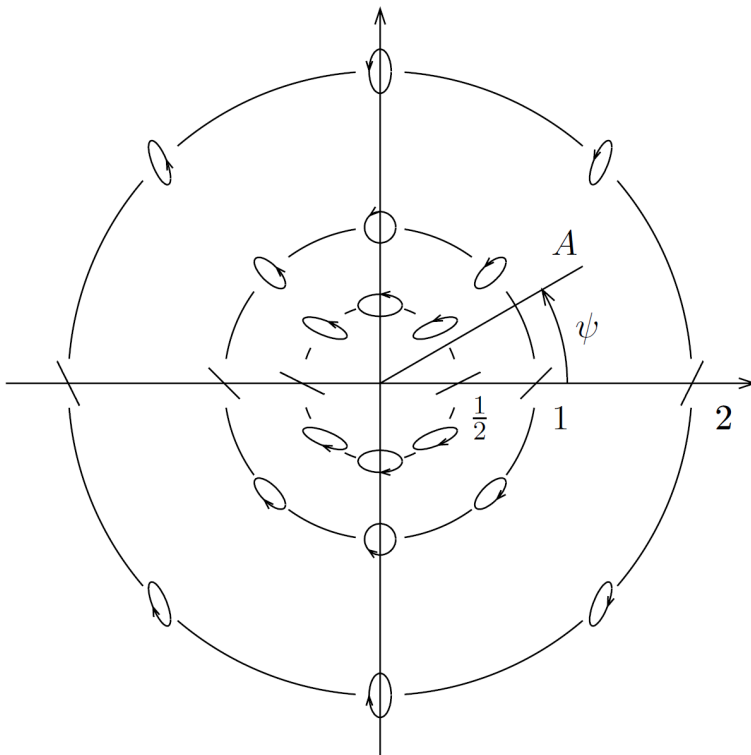


Figure 2: In this figure,  $\psi = -\alpha$  in our notes, and  $A = E_2/E_1$  (Courtesy of J.A. Kong, Electromagnetic Wave Theory).

Figure 3 shows a graphic picture of a CP wave propagating through space.

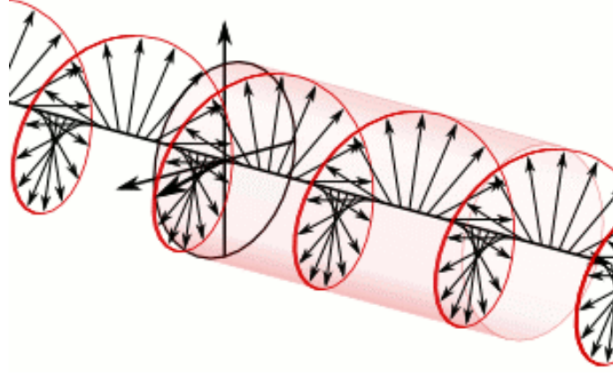


Figure 3: Courtesy of Wikipedia.

## 2.1 Arbitrary Polarization Case and Axial Ratio

The axial ratio (AR) is an important figure of merit for designing CP antennas (antennas that will radiate CP waves). The closer is this ratio to 1, the better the antenna design. We will discuss the general polarization and the axial ratio of a wave.

For the general case for arbitrary  $\alpha$ , we let

$$E_x = E_1 \cos \omega t, \quad E_y = E_2 \cos(\omega t + \alpha) = E_2(\cos \omega t \cos \alpha - \sin \omega t \sin \alpha) \quad (2.11)$$

Then from the above, expressing  $E_y$  in terms of  $E_x$ , one gets

$$E_y = \frac{E_2}{E_1} E_x \cos \alpha - E_2 \left[ 1 - \left( \frac{E_x}{E_1} \right)^2 \right]^{1/2} \sin \alpha \quad (2.12)$$

Rearranging and squaring, we get

$$aE_x^2 - bE_x^2 E_y^2 + cE_y^2 = 1 \quad (2.13)$$

where

$$a = \frac{1}{E_1^2 \sin^2 \alpha}, \quad b = \frac{2 \cos \alpha}{E_1 E_2 \sin \alpha}, \quad c = \frac{1}{E_2^2 \sin^2 \alpha} \quad (2.14)$$

After letting  $E_x \rightarrow x$ , and  $E_y \rightarrow y$ , equation (2.13) is of the form,

$$ax^2 - bxy + cy^2 = 1 \quad (2.15)$$

The equation of an ellipse in its self coordinates is

$$\left( \frac{x'}{A} \right)^2 + \left( \frac{y'}{B} \right)^2 = 1 \quad (2.16)$$

where  $A$  and  $B$  are axes of the ellipse as shown in Figure 4. We can transform the above back to the  $(x, y)$  coordinates by letting

$$x' = x \cos \theta - y \sin \theta \quad (2.17)$$

$$y' = x \sin \theta + y \cos \theta \quad (2.18)$$

to get

$$x^2 \left( \frac{\cos^2 \theta}{A^2} + \frac{\sin^2 \theta}{B^2} \right) - xy \sin 2\theta \left( \frac{1}{A^2} - \frac{1}{B^2} \right) + y^2 \left( \frac{\sin^2 \theta}{A^2} + \frac{\cos^2 \theta}{B^2} \right) = 1 \quad (2.19)$$

Comparing (2.13) and (2.19), one gets

$$\theta = \frac{1}{2} \tan^{-1} \left( \frac{2 \cos \alpha E_1 E_2}{E_2^2 - E_1^2} \right) \quad (2.20)$$

$$\text{AR} = \left( \frac{1 + \Delta}{1 - \Delta} \right)^{1/2} > 1 \quad (2.21)$$

where AR is the axial ratio where

$$\Delta = \left( 1 - \frac{4E_1^2 E_2^2 \sin^2 \alpha}{(E_1^2 + E_2^2)^2} \right)^{1/2} \quad (2.22)$$

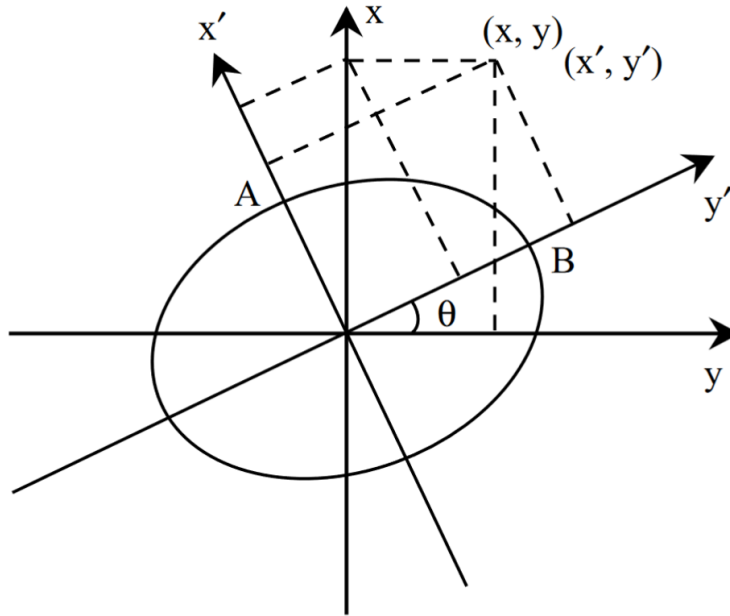


Figure 4:



### 3 Polarization and Power Flow

For a linearly polarized wave,

$$\mathbf{E} = \hat{x}E_0 \cos(\omega t - \beta z), \quad \mathbf{H} = \hat{y} \frac{E_0}{\eta} \cos(\omega t - \beta z) \quad (3.1)$$

Hence, the instantaneous power becomes

$$\mathbf{S}(t) = \mathbf{E}(t) \times \mathbf{H}(t) = \hat{z} \frac{E_0^2}{\eta} \cos^2(\omega t - \beta z) \quad (3.2)$$

indicating that for a linearly polarized wave, the instantaneous power is function of both time and space. It travels as lumps of energy through space. In the above  $E_0$  is the amplitude of the linearly polarized wave.

Next, we look at power flow for elliptically and circularly polarize waves. It is to be noted that in the phasor world or frequency domain, (2.1) becomes

$$\mathbf{E}(z, \omega) = \hat{x}E_1 e^{-j\beta z} + \hat{y}E_2 e^{-j\beta z + j\alpha} \quad (3.3)$$

For LHEP,

$$\mathbf{E}(z, \omega) = e^{-j\beta z} (\hat{x}E_1 + j\hat{y}E_2) \quad (3.4)$$

whereas for LHCP

$$\mathbf{E}(z, \omega) = e^{-j\beta z} E_1 (\hat{x} + j\hat{y}) \quad (3.5)$$

For RHEP, the above becomes

$$\mathbf{E}(z, \omega) = e^{-j\beta z} (\hat{x}E_1 - j\hat{y}E_2) \quad (3.6)$$

whereas for RHCP, it is

$$\mathbf{E}(z, \omega) = e^{-j\beta z} E_1 (\hat{x} - j\hat{y}) \quad (3.7)$$

Focussing on the circularly polarized wave,

$$\mathbf{E} = (\hat{x} \pm j\hat{y}) E_0 e^{-j\beta z} \quad (3.8)$$

Using that

$$\mathbf{H} = \frac{\boldsymbol{\beta} \times \mathbf{E}}{\omega \mu},$$

then

$$\mathbf{H} = (\mp \hat{x} - j\hat{y}) j \frac{E_0}{\eta} e^{-j\beta z} \quad (3.9)$$

Therefore,

$$\mathbf{E}(t) = \hat{x}E_0 \cos(\omega t - \beta z) \pm \hat{y}E_0 \sin(\omega t - \beta z) \quad (3.10)$$

$$\mathbf{H}(t) = \mp \hat{x} \frac{E_0}{\eta} \sin(\omega t - \beta z) + \hat{y} \frac{E_0}{\eta} \cos(\omega t - \beta z) \quad (3.11)$$

Then the instantaneous power becomes

$$\mathbf{S}(t) = \mathbf{E}(t) \times \mathbf{H}(t) = \hat{z} \frac{E_0^2}{\eta} \cos^2(\omega t - \beta z) + \hat{z} \frac{E_0^2}{\eta} \sin^2(\omega t - \beta z) = \hat{z} \frac{E_0^2}{\eta} \quad (3.12)$$

In other words, a CP wave delivers constant power independent of space and time.

It is to be noted that the complex Poynting vector

$$\mathbf{S} = \mathbf{E} \times \mathbf{H}^* \quad (3.13)$$

are real both for linearly, circularly, and elliptically polarized waves. This is because there is no reactive power in a plane wave of any polarization: the stored energy in the plane wave cannot be returned to the source!